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REAL TIME DIGITAL SIGNAL PROCESSING

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Frequency Analysis

Fast Fourier Transform (FFT)

Fast Fourier Transform

DFT:
$$X[k] = \sum_{n=0}^{N-1} x[n] W_N^{kn}, \qquad k = 0, 1, ..., N-1,$$

 N^2 complex multiplications N(N-1) complex aditions

$$X[k] = \sum_{n=0}^{N-1} \left[(\mathcal{R}e\{x[n]\}\mathcal{R}e\{W_N^{kn}\} - \mathcal{J}m\{x[n]\}\mathcal{J}m\{W_N^{kn}\}) + j(\mathcal{R}e\{x[n]\}\mathcal{J}m\{W_N^{kn}\} + \mathcal{J}m\{x[n]\}\mathcal{R}e\{W_N^{kn}\}) \right], \qquad 4N^2 \text{ real multiplications}$$
$$h = 0, 1, \dots, N-1, \qquad k = 0, 1, \dots, N-1,$$

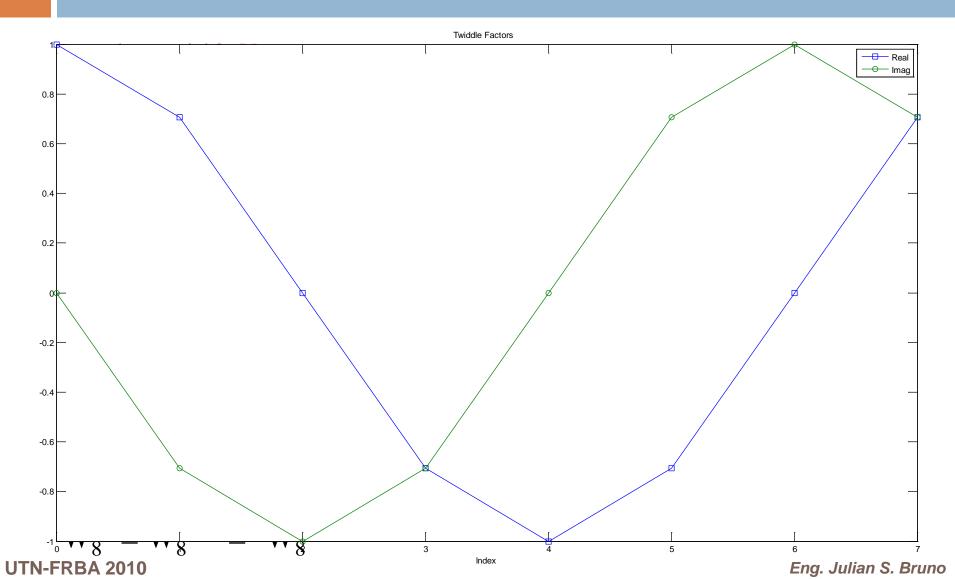
1.
$$W_N^{k[N-n]} = W_N^{-kn} = (W_N^{kn})^*$$
 (complex conjugate symmetry);
2. $W_N^{kn} = W_N^{k(n+N)} = W_N^{(k+N)n}$ (periodicity in *n* and *k*).

Computational algorithms that exploit both the symmetry and the periodicity of the sequence W_N^{kn} has come to be know as the fast Fourier transform, or FFT.

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N-1

Applying the properties of symmetry and periodicity to W_N^r for N=8



Decimation-In-Time FFT algorithms I

$$X[k] = \sum_{n \text{ even}} x[n] W_N^{nk} + \sum_{n \text{ odd}} x[n] W_N^{nk}$$

$$X[k] = \sum_{r=0}^{(N/2)^{-1}} x[2r] W_N^{2rk} + \sum_{r=0}^{(N/2)^{-1}} x[2r+1] W_N^{(2r+1)k}$$

$$= \sum_{r=0}^{(N/2)^{-1}} x[2r] (W_N^2)^{rk} + W_N^k \sum_{r=0}^{(N/2)^{-1}} x[2r+1] (W_N^2)^{rk}$$

$$X[k] = \sum_{r=0}^{(N/2)^{-1}} x[2r] W_{N/2}^{rk} + W_N^k \sum_{r=0}^{(N/2)^{-1}} x[2r+1] W_{N/2}^{rk}$$

$$X[k] = \sum_{r=0}^{(N/2)^{-1}} x[2r] W_{N/2}^{rk} + W_N^k \sum_{r=0}^{(N/2)^{-1}} x[2r+1] W_{N/2}^{rk}$$

$$x[1] \longrightarrow_{x[3]} W_N^{k} = e^{-2j(2\pi/N)} = e^{-j2\pi/(N/2)} = W_{N/2}$$

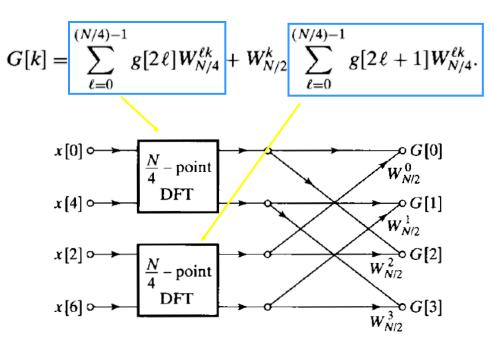
$$x[1] \longrightarrow_{x[3]} W_N^{k} = 0, 1, \dots, N-1.$$

$$x[n] \longrightarrow_{x[3]} W_N^{k} = 0, 1, \dots, N-1.$$

Decimation-In-Time FFT algorithms

$$G[k] = \sum_{r=0}^{(N/2)-1} g[r] W_{N/2}^{rk} = \sum_{\ell=0}^{(N/4)-1} g[2\ell] W_{N/2}^{2\ell k} + \sum_{\ell=0}^{(N/4)-1} g[2\ell+1] W_{N/2}^{(2\ell+1)k},$$

or



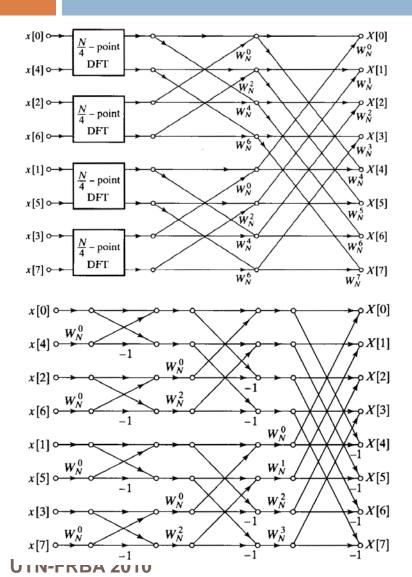
$$W_{N/2}^{k} = e^{-j2\pi k/(N/2)} = W_{N}^{2k}$$
$$W_{N/2}^{0} = W_{N}^{0}$$
$$W_{N/2}^{1} = W_{N}^{2}$$
$$W_{N/2}^{2} = W_{N}^{4} = -W_{N}^{0}$$
$$W_{N/2}^{3} = W_{N}^{6} = -W_{N}^{2}$$

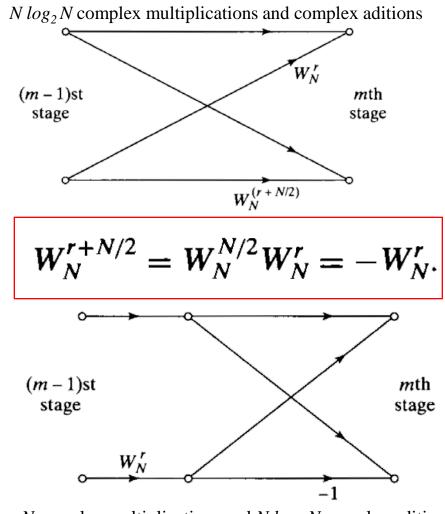
Similarly, H[k] would be represented as

$$H[k] = \sum_{\ell=0}^{(N/4)-1} h[2\ell] W_{N/4}^{\ell k} + W_{N/2}^{k} \sum_{\ell=0}^{(N/4)-1} h[2\ell+1] W_{N/4}^{\ell k}$$

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Decimation-In-Time FFT algorithms





 $N/2 \log_2 N$ complex multiplications and $N \log_2 N$ complex aditions Eng. Julian S. Bruno

FFT vs. DFT

The FFT is simply an algorithm for efficiently calculating the DFT

Computational efficiency of an N-Point FFT:

DFT: N² Complex Multiplications
 FFT: (N/2) log₂(N) Complex Multiplications

N	DFT Multiplications	FFT Multiplications	FFT Efficiency
256	65,536	1,024	64 : 1
512	262,144	2,304	114 : 1
1,024	1,048,576	5,120	205 : 1
2,048	4,194,304	11,264	372 : 1
4,096	16,777,216	24,576	683 : 1

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Bit Reversal

Decimal Number :	0	1	2	3	4	5	6	7
Binary Equivalent :	000	001	010	011	100	101	110	111
Bit-Reversed Binary :	000	100	010	110	001	101	011	111
Decimal Equivalent :	0	4	2	6	1	5	3	7

□ The bit reversal algorithm used to perform the re-ordering of signals.

- The decimal index, n, is converted to its binary equivalent.
- The binary bits are then placed in reverse order, and converted back to a decimal number.
- Bit reversing is often performed in DSP hardware in the data address generator (DAG).

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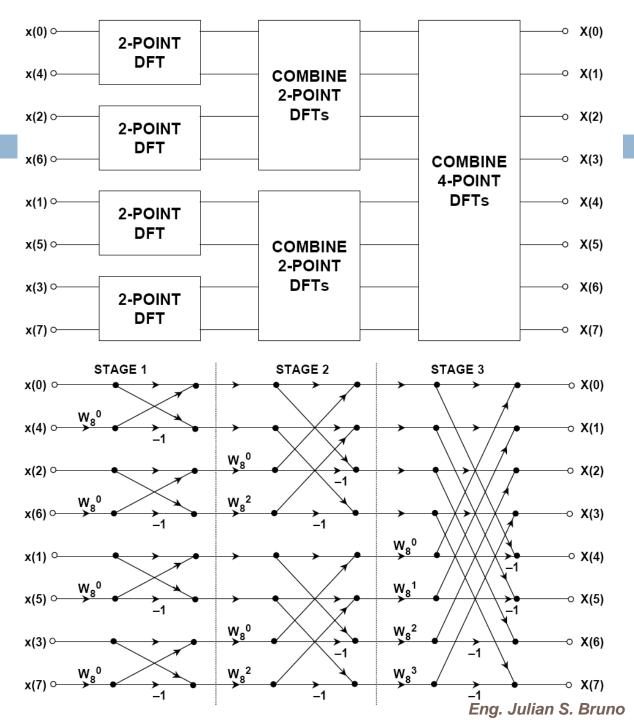
DIT FFT

 Input signal must be properly re-ordered using a *bit reversal* algorithm

- In-place computation
- Number of stages: *log*₂ *N*
- Stage 1: all the twiddle factors are 1
- Last Stage: the twiddle factors are in sequential order

	Stage 1	Stage 2	Stage 3	Stage Log ₂ N
Number of Groups	N/2	N/4	N/8	1
Butterflies per Group	1	2	4	N/2
Dual-Node Spacing	1	2	4	N/2
Twiddle Factor Exponents	(N/2)k, k=0	(N/4)k, k=0,1	(N/8)k, k=0,1, 2,3	k, k=0 to N/2–1

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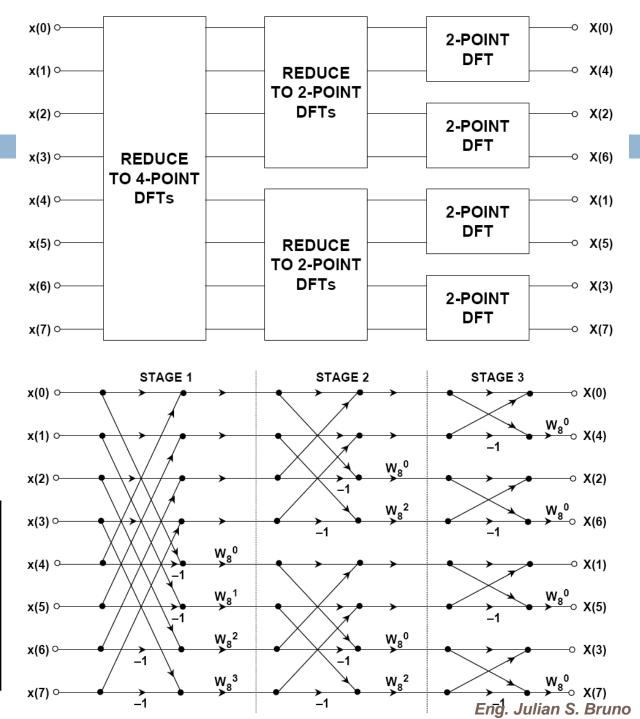
DIF FFT

 Output signal must be properly re-ordered using a *bit reversal* algorithm

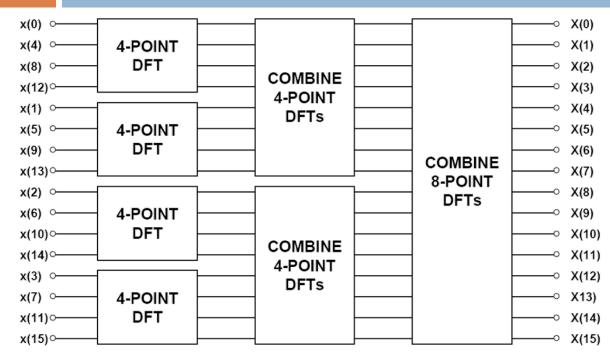
- In-place computation
- Number of stages: log₂ N
- Stage 1: the twiddle factors are in sequential order
- Last Stage: all the twiddle factors are 1

	Stage 1	Stage 2	Stage 3	Stage Log ₂ N
Number of Groups	1	2	4	N/2
Butterflies per Group	N/2	N/4	N/8	1
Dual-Node Spacing	N/2	N/4	N/8	1
Twiddle Factor Exponents	n, n=0 to N/2 - 1	2n, n=0 to N/4 - 1	4n, n=0 to N/8 - 1	(N/2)n, n=0

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Radix-4 Decimation-In-Time FFT Algorithm



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•A radix-4 FFT combines two stages of a radix-2 FFT into one, so that half as many stages are required.

•The radix-4 butterfly is consequently larger and more complicated than a radix-2 butterfly.

•N/4 butterflies are used in each of $(\log_2 N)/2$ stages, which is one quarter the number of butterflies in a radix-2 FFT.

•Addressing of data and twiddle factors is more complex, a radix-4 FFT requires fewer calculations than a radix-2 FFT.

It can compute a radix-4 FFT significantly faster than a radix-2 FFT

Hardware benchmark comparisons

- **ADSP-2189M**, 16-bit, Fixed-Point @ 75MHz
 - 453µs (1024-Point)

□ ADSP-21160 SHARC[™], 32-bit, Floating-Point @ 100MHz

- 180µs (1024-Point), 2 channels, SIMD Mode
- 115µs (1024-Point), 1 channel, SIMD Mode

□ ADSP-TS001 TigerSHARCTM @ 150MHz,

- 16-bit, Fixed-Point Mode
 - 7.3µs (256-Point FFT)
- **32-bit, Floating-Point Mode**
 - 69µs (1024-Point)

Real Time FFT considerations

- Signal Bandwidth
- Sampling Frequency, fs
- Number of Points in FFT, N
- Frequency Resolution = fs/N
- Maximum Time to Calculate N-Point FFT = N/fs
- Fixed-Point vs. Floating Point DSP
- Radix-2 vs. Radix-4 Execution Time
- Windowing Requirements

Implementation DIT FFT in ADSP 2181 First Stage

```
W_{N} = e^{-j2\pi/N} \cos(2\pi/N) - j\sin(2\pi/N)
                                                              x_`+jy_
                                       x<sub>0</sub> + jy<sub>0</sub>
                                                                      W_N = C + i(-S)
                                                                     A = (C) x_1 - (-S) y_1
          i0 = inplacereal;
                                            W=C+i(-S)
                                                                     B = (C) y_1 + (-S) x_1
                                                             x<sub>1</sub>' + jy <sub>1</sub>
          i1 = inplacereal + 1;
                                       x₁ + jy ₁—⊗-
          i2 = inplaceimag;
                                                                     x_0' = x_0 + A y_0' = y_0 + B
          i3 = inplaceimag + 1;
                                                                     x_1 = x_0 - A \quad y_1 = y_0 - B
                                                A+iB
          m^2 = 2;
                                                      //cntr=N/2
          cntr = nover2;
          ax0 = dm(i0, m0);
                                                      //ax0=x0
                                                      //av0=x1
          ay0 = dm(i1, m0);
          ay1 = dm(i3, m0);
                                                      //av1=v1
          do group lp until ce;
                 ar=ax0+ay0, ax1=dm(12,m0);
                                                     //ar=x0+x1, ax1=y0
                 sb=expadj ar, dm(10,m2)=ar;
                                                      //Check for bit growth, x0'=x0+x1
                 ar=ax0-ay0;
                                                      //ar=x0-x1
                 sb=expadj ar;
                                                      //Check for bit growth
                 dm(i1,m2)=ar, ar=ax1+ay1;
                                                     //x1'=x0-x1, ar=y0+y1
                                                     //Check for bit growth, y0'=v0+v1
                 sb=expadj ar, dm(12,m2)=ar;
                 ar=ax1-ay1, ax0=dm(10,m0);
                                                     //ar=y0-y1, ax0= next x0
                 sb=expadj ar, dm(i3,m2)=ar;
                                                     //Check for bit growth, y1'=y0-y1
                                                     //ay0= next x1
                 ay0=dm(i1,m0);
                 ay1=dm(i3,m0);
                                                      //av0= next v1
group lp:
          call bfp adj;
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```

Implementation DIT FFT in ADSP 2181 Butterfly Loop

```
I4=twid_real;
I5=twid_imag;
CNTR=DM(bflys_per_group);
MY0=PM(I4,M4),MX0=DM(I1,M0);
MY1=PM(I5,M4),MX1=DM(I3,M0);
```

```
DO bfly_loop UNTIL CE;
MR=MX0*MY1(SS),AX0=DM(I0,M0);
MR=MR+MX1*MY0(RND),AX1=DM(I2,M0);
AY1=MR1,MR=MX0*MY0(SS);
MR=MR-MX1*MY1(RND);
AY0=MR1,AR=AX1-AY1;
```

SB=EXPADJ AR, DM(I3, M1)=AR;

AR=AXO-AYO, MX1=DM(I3, MO), MY1=PM(I5, M4); //AR=xO-[x1(C)-y1(-S)],

SB=EXPADJ AR, DM(I1, M1)=AR;

AR=AX0+AY0, MX0=DM(I1, M0), MY0=PM(I4, M4); //AR=x0+[x1(C)-y1(-S)],

SB=EXPADJ AR, DM(IO, M1)=AR;

AR=AX1+AY1; bfly loop: SB=EXPADJ AR, DM(I2, M1)=AR; //I4 --> C of W0
//I5 --> (-S) of W0
//CNTR = butterfly counter
//MY0=C, MX0=x1
//MY1=-S, MX1=y1

```
//MR=x1(-S), AXO=xO
//MR=(y1(C)+x1(-S)),AX1=y0
//AY1=y1(C)+x1(-S), MR=x1(C)
//MR=x1(C)-y1(-S)
//AYO=x1(C)-y1(-S),
//AR=y0-[y1(C)+x1(-S)]
//Check for bit growth,
//y1'=y0-[y1(C)+x1(-S)]
//MX1=next y1, MY1=next (-S)
//Check for bit growth,
//x1'=x0-[x1(C)-y1(-S)]
//MXO=next x1, MYO=next C
//Check for bit growth,
//x0' = x0 + [x1(C) - y1(-S)]
//AR=y0+[y1(C)+x1(-S)]
//Check for bit growth,
//y0'=y0+[y1(C)+x1(-S)]
```

Implementation DIT FFT in ADSP 2181 **Block Floating-Point Scaling Routine**

bfp_adj:		$A = (C) x_1 - (-S) y_1$
AY0=CNTR;	//{Check for last stage}	$B = (C) y_1 + (-S) x_1$
AR=AY0-1;		$x_0' = x_0 + A$
IF EQ RTS;	<pre>//{If last stage, return}</pre>	$y_0' = y_0 + B$
AY0=-2;		$x_1' = x_0 - A$
AX0=SB;		1 0
AR=AX0-AY0;	//{Check for SB=-2}	$y_1 = y_0 - B$
IF EQ RTS;	<pre>//{IF SB=-2, no bit growth, return} //{IF speed pointan}</pre>	
IO=inplacereal; I1=inplacereal;	<pre>//{I0=read pointer} //{I1=write pointer}</pre>	$x_0' < 1$, $y_0' < 1$
AY0=-1;	//{II-write pointer}	0 00
MY0=0x4000;	//{Set MYO to shift 1 bit right}	$ C_{max} = 1$, $ S_{max} = 1$
AR=AXO-AYO, MXO=DM(I0,M1);	<pre>//{Check if SB=-1; Get first sample}</pre>	
IF EQ JUMP strt shift;	<pre>//{If SB=-1, shift block data 1 bit}</pre>	$x_0' = x_0 + x_1 + y_1 < 1$
AY0=-2;	<pre>//{Set AY0 for block exponent update}</pre>	$x_0 < 0.33$, $x_1 < 0.33$, $y_1 < 0.33$
MY0=0x2000;	//{Set MY0 to shift 2 bits right}	
strt_shift:		$y_0' = y_0 + y_1 - x_1 < 1$
	(((initialize lease souther))	
CNTR=2047;	//{initialize loop counter}	
DO shift_loop UNTIL CE;	<pre>//{Shift block of data} //{MR=shifted data,MX0=next value}</pre>	0.33 = 0x2A3D
<pre>shift_loop: DM(I1,M1)=MR1;</pre>	//{Unshifted data=shifted data}	$0.00 = 0 \times 2 \times 0 D$
SHITC_TOOP. DM(II,MI)-MKI,	// [UNSHITCED Data-SHITCED Data]	
MR=MX0*MY0(RND);	//{Shift last data word}	0.25
AY0=DM(blk exponent);	//{Update block exponent and}	0.25
DM(I1,M1)=MR1,AR=AY0-AX0;	//{store last shifted sample}	
DM(blk_exponent)=AR;		
sb = -2;	-2^{0} 2^{-1} 2^{-2} 2^{-3} 2^{-4} 2^{-5} 2^{-6} 2^{-7} 2^{-8} 2^{-9} 2^{-1}	0 2-11 2-12 2-13 2-14 2-15
RTS;	$\begin{vmatrix} -2^{0} & 2^{-1} & 2^{-2} & 2^{-3} & 2^{-4} & 2^{-5} & 2^{-6} & 2^{-7} & 2^{-8} & 2^{-9} & 2^{-1} \end{vmatrix}$	
	0 0 1 0 X X X X X X X X	XXXXX
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Implementation DIT FFT in ADSP 2181 Scramble Routine

scramble:

```
I4=inputreal;
I0=inplacereal;
M4=1;
M0=mod_value;
L4=0;
L0=0;
cntr = N;
ENA BIT_REV;
D0 brev UNTIL CE;
AY1=DM(I4,M4);
DM(I0,M0)=AY1;
DIS BIT_REV;
RTS;
scramble.end;
```

```
//{I4-->sequentially ordered data}
//{I0-->scrambled data}
```

```
//{MO=modifier for reversing N bits}
```

//{Enable bit-reversed outputs on DAG1}

```
//{Read sequentially ordered data}
//{Write data in bit-reversed location}
//{Disable bit-reverse}
//{Return to calling program}
```